

***Remarks***

The Action objects to Figs. 9-12 as for failing to include the designation "Prior Art." Figs. 9 and 11 do not constitute prior art with respect to the present invention, and thus, have not been labeled as such. Figs. 10 and 12 have been labeled "Prior Art."

Submitted herewith is a substitute specification to address the objections in the Action. No new matter has been added. A copy of the specification showing the changes to the originally filed specification is also attached.

The Action rejected claims 1-17 under 35 U.S.C. §112, second paragraph. Claims 1-17 have been canceled and replaced by new claims 18-27. The new claims comply with Section 112.

Claims 1-7 and 10-17 stand rejected under 35 U.S.C. §103(a) as being unpatentable over Ganz in view of what is characterized by the Action as applicant's admitted prior art shown in Fig. 9. Claims 1-17 have been canceled and replaced by claims 18-27.

As discussed on page 3 of the present specification, Fig. 9 shows the transformer disclosed in IEEE Transactions on Electromagnetic Compatibility, Vol. 41, No. 3, August 1999. This article was authored by the inventor of the present application, and was published less than one year prior to the priority date. Accordingly, the article would not constitute prior art under any of the sections of 35 U.S.C. §102 or §103. The specification characterizes the subject matter disclosed in this article and shown in Fig. 9 as "Background," but does not state that it constitutes prior art under the Patent Statute. Accordingly, reconsideration of the rejection is respectfully requested.

Even if Fig. 9 were considered prior art with respect to the present invention, the combination

of Fig. 9 and Ganz would not render claims 18-27 obvious. As discussed in col. 1, lines 13-16 of Ganz, “[a]n object of the invention is to increase the ratio of the series aiding impedance to the parallel opposing impedance of the windings of inductance devices.” Ganz teaches that “inductance 20 comprises two windings 22 and 23 wound on shell type core 25. Strips of conductive material, such as copper foil, 26, 27, 28, 29 and 30, are disposed between adjacent layers of then windings while paper strips may also be placed between the layers in the usual manner if desirable. The ends of each conductive strip are either insulated or spaced from one after another in order to prevent formation of short-circuit turns.” (col. 3, lines 33-42). As further discussed in col. 4, lines 48-55 Ganz states, “[t]he invention is more effective at carrier and higher frequencies because of the greater shielding efficiency of the conductive strips at these higher frequencies. It has been observed, further, the effectiveness may be increased somewhat by increasing the number of layers of windings or by increasing the thickness of the conductive strips.” Accordingly, from the above description, the invention disclosed in Ganz is an inductance device and does not constitute an isolation transformer as in the present invention and shown in Fig. 9. Accordingly, even if Ganz does teach the elements for which it is cited for, it would not be obvious to modify it with the teachings of Fig. 9.

Furthermore, the Action argues that “the conductive strip” of Ganz is analogous to the “short-circuit ring” of the claimed invention, except for the “specific structure.” However, the conductive strip of Ganz has both ends insulated or spaced from one another so as not to form a short-circuit ring and states that it is also desirable to make the strips as thick as possible.

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Accordingly, "the conductive strip" of Ganz completely different and not analogous from "the short-circuit ring having a wide surface area" of the present invention.

Claims 8 and 9 stand rejected under 35 U.S.C. §103(a) as being unpatentable over Sakaguchi. Canceled claims 8 and 9 have not been replaced by corresponding new claims. In any event, Sakaguchi does not teach the short circuit rings of claims 18-27.

Based on the above Amendment and the following Remarks, Applicants respectfully request that the Examiner reconsider all outstanding objections and rejections and that they be withdrawn. A Notice of Allowance with claims 18-27 is respectfully requested.

Respectfully submitted,

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**[SUBSTITUTE SPECIFICATION]**  
**[DOCUMENT] — SPECIFICATION**

**[TITLE OF INVENTION]**

Isolation Transformers

[CLAIMS]

~~[CLAIM 1] An isolation transformer comprised by a multi layer, multi-winding primary coil, a multi layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil functions as an isolation (noise cutoff) transformer by changing the coil layers of one or both of the coils formed by winding an insulated, covered, copper wire to a multi layer, multi-winding coil, into which a number of short circuit rings made of conducting films are inserted and layered,~~

and

~~the surface area of the aforementioned conducting short circuit rings is made approximately as large as that of the neighboring coil layers, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high frequency region, where resonances should be suppressed.~~

~~[CLAIM 2] An isolation transformer comprised by a multi layer, multi-winding primary coil, a multi layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil, functions as an isolation transformer by changing the coil layers of one or both of the coils formed by~~

~~winding spirally an insulated, covered, copper wire to a multi-layer, multi-winding coil, into which a number of short circuit rings made of conducting films are inserted and layered, and the planer configuration of the aforementioned conducting short circuit rings is made approximately identical to that of the neighboring coil layers, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high frequency region, where resonances should be suppressed.~~

~~[CLAIM 3] An isolation transformer comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil functions as an isolation transformer by changing the coil layers of one or both of the coils formed by winding cylindrically an insulated, covered, copper wire to a multi-layer, multi-winding coil, into which a number of short circuit rings made of conducting films are inserted and layered,~~

~~and~~

~~the inner surface area of the aforementioned cylindrical short circuit rings is made approximately identical to the outer surface of the neighboring coil, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high-frequency region, where resonances should be suppressed.~~

~~[CLAIM 4] An Isolation transformer of claim 1, 2, or 3 characterized by the aforementioned short circuit ring inserted into every inter-coil layer space.~~

~~[CLAIM 5] An Isolation transformer of claim 1, 2, or 3 characterized by the  
aforementioned short circuit ring inserted into selected inter coil layer  
space.~~

~~[CLAIM 6] An Isolation transformer of claim 1, 2, or 3 characterized by the  
aforementioned short circuit rings made by laminating plastic films.~~

~~[CLAIM 7] An Isolation transformer of claim 1, 2, or 3 characterized by the  
aforementioned short circuit rings of thickness of 7  $\mu$ m or less.~~

~~[CLAIM 8] A transformer comprised by a multi layer, multi winding primary  
coil, a multi layer, multi winding secondary coil, and a core that forms a  
magnetic path between the aforementioned primary coil and the  
aforementioned secondary coil functions as an isolation transformer by  
changing the coil layers of one or both of the coils formed by winding an  
insulated, covered, copper wire to a multi layer, multi winding coil, and  
each layer of the multi layer, multi winding coil is formed by winding an  
insulated, covered, copper wire, the surface of which is further covered with  
a conducting film that is made approximately as thick as or less thicker  
than the skin depth of the induced current generated by the skin effect in  
the high frequency region, where resonances should be suppressed.~~

~~[CLAIM 9] An Isolation transformer of claim 8 characterized by the  
aforementioned conducting thin film of thickness of 7  $\mu$ m or less.~~

~~[DETAILED DESCRIPTION OF THE INVENTION]~~

~~[0001]~~

~~[TECHNICAL FIELD TO WHICH THE INVENTION BELONGS]~~

## Background of the Invention

This invention is related to isolation transformers that suppress high-frequency electromagnetic noise (hereafter called noise) transmitted through power transmission lines and/or signal transmission lines.

{0002}

~~{DESCRIPTION OF THE PRIOR ART}~~

Micro-computers are being used in various fields such as information, communication, and other industries in addition to daily life etc. This is due to down-sizing, lower prices, and higher reliability of micro-computers driven by the continuous advancement of integrated circuits. However, integrated circuits are controlled by extremely small electric energy, and therefore, are subject to errors and damages caused by some noise transmitted from external sources. Such events will eventually lead to various inconveniences and/or accidents owing to malfunctioning or stopping of various equipment and devices that contain integrated circuits and of systems using them. Consequently, protection of electronic devices and equipment that contain densely packed and complicated circuits from noise-related troubles have become an urgent issue.

{0003}

For suppression of noise-related troubles isolation transformers of electromagnetic-shield type have been used. The isolation transformers of electromagnetic-shield type have primary- and secondary coils isolated by approximately 20  $\mu m$ -thick aluminum foils. The isolation transformers of

electromagnetic-shield type have such attenuation characteristics of normal-mode noise as depicted in Figure 9. Namely, in the frequency range of several hundred Hz to 1 *MHz* the attenuation increases generally mildly with the frequency to -50*dB*. In the range from 1 *MHz* to 100*MHz* it takes the form of an irregular saw-tooth wave, which is comprised by troughs and crests of various sizes between the maximum of -78*dB* and the minimum of -24*dB*.

{0004}

The noise attenuation characteristic that occurs in the high-frequency region above a few *MHz*, and takes the form of an irregular saw-tooth wave, which is comprised by troughs and crests of various sizes is caused by stray capacitance appearing uniquely in each transformer; its origin can be traced to the multi-layer, multi-winding coils having numerous, irregular resonance circuits caused by complicated combinations of infinitesimally different inter-layer and inter-winding distributed capacitance as well as leakage inductances. Components with extremely complicated combinations such as multi-layer, multi-winding coils in transformers thus have random and complicated noise attenuation characteristics; as the crests have extremely low attenuation, this scheme cannot provide highly reliable isolation transformers. In order to improve the reliability of isolation transformers it is necessary to increase the attenuation in the high frequency region exceeding a few *MHz* and to fully decrease each amplitude of the saw-tooth wave comprised by troughs and crests of various sizes, suppressing the crests. As the irregular curves of the attenuation characteristic is unique to

each transformer, the ideal suppression mechanism ought to commonly and effectively apply to any transformer. However, it has been impossible for the isolation transformers of electromagnetic-shield type to satisfy this condition.

{0005}

Therefore, this inventor has developed two types of isolation transformers to resolve the aforementioned issue confronting the isolation transformers of electromagnetic-shield type: One of them was made public in the Japanese Patent No. 2,645,256, and, as shown in Figure 10, it is an isolation transformer, which is characterized by shielding entirely both primary- and secondary coils by short-circuit rings 4 made of conducting thin films

- of thickness of  $0.5 \sim 100 \mu m$ .

{0006}

Another is the type published in IEEE (U.S. Institute of Electrical and Electronics Engineers) Transactions on Electromagnetic Compatibility, Vol. 41, No. 3, August 1999. As shown in Figure 9, this is an isolation transformer characterized by positioning in the vicinity of, or more specifically, in-between primary- and secondary coils a short-circuit ring 4 made of a conducting thin film, which has a thickness of approximately  $7 \mu m$  or less (hereafter abbreviated as isolation transformer of short-circuit-ring type). As shown in Figure 6 as an example, the core that forms a magnetic path between the primary coil 1 and the secondary coil 2, is

manufactured by stacking together pieces of type E core and pieces of type I core of specified dimensions to specified thickness after stamping them out of an isotropic silicon steel sheet of thickness of  $0.5\text{mm}$ . Furthermore, the short-circuit ring 4 of conducting thin film is fabricated, as shown in Figure 5 for an example, by cutting a ring, which is approximately as thick as the primary coil 1 and secondary coil 2, out of a rolled aluminum foil of thickness of  $7\mu\text{m}$  and by laminating it to a durable polyester film of thickness of  $50\mu\text{m}$ .

{0007}

This short-circuit ring 4 made of metallic thin film with a large surface area functions as the tertiary coil connected to the primary- and secondary coils, respectively. Through this short -circuit ring 4 of conducting thin film flows the current induced by fundamental wave current flowing through the primary coil, its higher harmonics, and high-frequency noise from external sources. In this case, as the high-frequency components flow essentially at the surface of the conductor only owing to the skin effect, they flow around the short-circuit ring, even if the ring is thin. Therefore, being effectively attenuated by the resistivity of the short-circuit ring, it is hard for the high-frequency components to flow from the primary coil to the secondary coil. At the same time, since the resistivity of the short-circuit ring functions as a resistance commonly inserted to all of the resonance circuits, which are distributed in the form of a number of infinitesimally small irregularly present resonance circuits due to complex combinations of capacitance and leakage inductances irregularly distributed in the coils, the resistivity of the

short-circuit ring has the effect of dramatically reducing the amplitudes of resonances. To sum up, in the isolation transformers of short-circuit ring, a short-circuit ring of conducting thin film or a short-circuit ring of conducting thin film laminated with a plastic film is used and the surface area of the aforementioned short-circuit rings is made approximately as large as that of the neighboring coil layers, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high-frequency region, in which resonances should be suppressed.

{0008}

Meanwhile, the currents induced by the fundamental wave, which are low-frequency components, are reduced in proportion to the cross sectional area of the conducting thin film of the short-circuit ring 4. However, as the short-circuit ring is made of a thin film of thickness of  $7\ \mu\text{m}$ , its cross section is small albeit its width. Therefore, the induced currents of fundamental-wave component that flows through the short-circuit ring is extremely small. As a consequence, by positioning this short-circuit ring 4 with a wide surface area in the vicinity of the primary- and secondary coils, respectively, an isolation transformer of short-circuit-ring type, which can eliminate or filter high-frequency noise, while keeping the loss of the fundamental-wave negligibly small, is provided.

{0009}

Presented in Figure 11 is an example of attenuation characteristics for the normal-mode noise of the isolation transformer of short-circuit ring type depicted in

Figure 9. Namely, in the range from a few hundred *Hz* to 1*MHz* the attenuation increases generally mildly with the frequency to -60*dB*, while in the range from 1*MHz* to 100*MHz* it has the form of an irregular saw-tooth wave, which is comprised by crests and troughs of various sizes between the maximum of -100*dB* and the minimum of -53*dB*. Furthermore, in the range from 100*MHz* to 300*MHz* it has the form of an irregular saw-tooth wave, which is comprised by crests and troughs of various sizes between the maximum of -72*dB* and the minimum of -50*dB*.

{0010}

As is evident in Figure 11, the attenuation characteristic curve of the isolation transformer of short-circuit ring type has a relatively flat portion with steep crests and troughs replaced with crests and troughs of smaller amplitudes. In comparison with isolation transformers of electromagnetic-shield type the isolation transformers of short-circuit-ring type show considerable improvements concerning the attenuation characteristic of normal-mode noise in the high-frequency region over 1*MHz*. More specifically, as shown in Figure 12, the isolation transformer of electromagnetic-shield type has the lowest attenuation of -24*dB*, while as shown in Figure 11 the isolation transformer of short-circuit-ring type has that of -53*dB*, showing an outstanding improvement of 29*dB*. The same trend exists for the highest points of attenuation, which is -78*dB* for the isolation

transformer of electromagnetic-shield type, while  $-100dB$  for the isolation transformer of short-circuit-ring type, showing another outstanding improvement of  $22dB$ .

{0011}

Furthermore, in the high-frequency range over  $10MHz$  considerable improvements are observed in the normal-noise characteristic. Namely, as evident in the region encompassed by thick dotted lines, the normal-mode attenuation characteristic in the range from  $10MHz$  to  $100MHz$  of the isolation transformer of the electromagnetic-shield type has the highest point of  $-78dB$  and the lowest point of  $-40dB$ , while that of the isolation transformer of the short-circuit-ring type has the highest point of  $-91dB$  and the lowest point of  $-53dB$ , showing considerable improvements of  $13dB$  at both the highest and lowest attenuation points.

{0012}

Though not illustrated, the same trend exists for common-mode noise; isolation transformers of short-circuit-ring type show considerable improvements over isolation transformers of electromagnetic-shield type in the high-frequency region above a few  $MHz$ .

{0013}

Though inside such components having extremely complicated combinations as multi-layer, multi-winding coils, there exist a number of resonance circuits caused by complicated combinations of infinitesimally

different inter-layer and inter-winding distributed capacitance as well as leakage inductances, inside isolation transformers of short-circuit-ring type the effects of stray capacitance due to such resonance circuits are evidently reduced. Moreover, as in addition to increasing the attenuation rates in the high-frequency range above a few *MHz* the amplitudes of the irregular saw-tooth shaped waves with crests and troughs of various sizes were suppressed to the lowest possible level, the reliability of isolation transformers of the short-circuit-ring type has been considerably improved.

{0014}

However, as evidently shown in Figure 11, the suppression of amplitudes of the characteristic curve for the attenuation rate of normal-mode noise with irregular saw-tooth shaped waves with crests and troughs of various sizes in the high-frequency region above a few *MHz* is not sufficient yet. Therefore, there still remains a question with regard to the reliability of the conventional isolation transformers of short-circuit-ring type with a wide short-circuit ring of conducting thin film covering each surface of primary- and secondary coils and of the isolation transformers of short-circuit-ring type with a wide short-circuit ring of a conducting thin film positioned between and in the vicinity of primary- and secondary coils.

{0015}

~~{THE ISSUE TO BE RESOLVED BY THE INVENTION}~~

Brief Summary of the Invention

The issue to be resolved by the present invention is to provide isolation transformers with high noise attenuation rates as well as high reliability by sufficiently suppressing the amplitudes of noise attenuation characteristic curves, which are irregular, saw-tooth shaped waves with crests and troughs of various sizes, of multi-layer, multi-winding transformers,

~~{HOW TO RESOLVE THE ISSUE}~~

~~{0016}~~

An isolation transformer comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil, functions as an isolation transformer to resolve the aforementioned issue by changing the coil layers of one or both of the coils formed by winding an insulated, covered, copper-wire to a multi-layer, multi-winding coil, into which a number of short-circuit rings made of conducting films are inserted and layered.

The ~~planer~~planar configuration of the aforementioned conducting short-circuit rings is made approximately identical to that of the neighboring coil-layers, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high-frequency region, in which resonances should be suppressed. The aforementioned short-circuit ring is inserted in between every coil layer or in-between selected coil layers.

As the aforementioned short-circuit ring, a short-circuit ring of conducting thin film or a short-circuit ring of conducting thin film laminated with a plastic film is used.

The aforementioned short-circuit ring has a thickness of  $7\ \mu\text{m}$  or less.

{0017}

Also, an isolation transformer comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil functions as an isolation transformer to resolve the aforementioned issue by changing the coil layers of one or both of the coils formed by winding spirally an insulated, covered, copper-wire to a multi-layer, multi-winding coil, into which a number of short-circuit rings made of conducting film are inserted and layered.

The ~~planer~~planar configuration of the aforementioned conducting short-circuit rings is made approximately identical to that of the neighboring coil-layers, and its thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high-frequency region, in which resonances should be suppressed.

The aforementioned short-circuit rings are inserted into every coil layer or into selected coil layers. As the aforementioned short-circuit rings, short-circuit rings of conducting thin film or short-circuit rings of conducting thin film laminated with a plastic films are used.

The aforementioned short-circuit rings have a thickness of  $7\ \mu\text{m}$  or less.

{0018}

Moreover, an isolation transformer comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil functions as an isolation transformer to resolve the aforementioned issue by changing the coil layers of one or both of the coils formed by winding cylindrically an insulated, covered, copper-wire to a multi-layer, multi-winding coil, into which a number of short-circuit rings made of conducting film are inserted and layered.

Furthermore, the inner surface of the aforementioned cylindrical short-circuit rings is made approximately identical to the outer surface of the neighboring coil, and their thickness is made approximately identical to or less than the skin depth of the induced current generated by the skin effect in the high-frequency region, in which resonances should be suppressed.

The aforementioned short-circuit rings are inserted in-between every coil layer or in-between selected coil layers.

As the aforementioned short-circuit rings, short-circuit rings of conducting thin film or short-circuit rings of conducting thin films laminated with plastic films are used. The aforementioned short-circuit rings have a thickness of  $7\ \mu m$  or less.

{0019}

Furthermore, an isolation transformer comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a core that forms a magnetic path between the aforementioned primary coil and the aforementioned secondary coil functions as an isolation transformer by changing the coil layers of one or both of the coils formed by winding an insulated, covered, copper-wire to a multi-layer, multi-winding coil.

Each layer of the multi-layer, multi-winding coils are formed by winding an insulated, covered, copper-wire, the surface of which is further covered with a conducting film that is made approximately as thick as or less thicker than the skin depth of the induced current generated by the skin effect in the high-frequency region, where resonances should be suppressed.

The aforementioned conducting thin films have a thickness of  $7\text{ }\mu\text{m}$  or less.

#### Brief Description of the Drawings

Figure 1 is a cross section view of the isolation transformer in the first embodiment.

Figure 2 is a cross section view of magnified part of the isolation transformer in the first embodiment shown in FIGURE 1.

Figure 3 is a cross section view of the isolation transformer of a modification in the first embodiment

Figure 4 is a cross section of the isolation transformer of another modification in the first embodiment

Figure 5 is a plan view of an example of circular, short-circuit ring of conducting thin film

Figure 6 is a oblique view of an example of the core.

Figure 7 is a cross section of the isolation transformer in the second embodiment of the present invention

Figure 8 is a cross section view of magnified part of the isolation transformer in the second embodiment shown in FIGURE 7.

Figure 9 is a cross section of an example of existing isolation transformer of short-circuit ring.

Figure 10 is a cross section of another example of ~~conventional~~ isolation transformer of short-circuit ring type.

Figure 11 shows a characteristic curve of attenuation rate for normal mode noise of a ~~conventional~~ isolation transformer of short-circuit ring type.

Figure 12 shows a characteristic curve of attenuation rate for normal mode noise of a ~~conventional~~ isolation transformer of electromagnetic-shield type.

{0020}

{DESCRIPTION OF THE PREFERRED EMBODIMENTS}

#### Detailed Description of the Invention

Figure 1 depicts a cross section of an isolation transformer of the short-circuit-ring type, which is the first embodiment of the present invention with its bobbin and core omitted and with much reduced numbers of windings and layers for easier comprehension. Figure 2 depicts a magnified part of Figure 1. The primary coil is a ring-type coil, in which an

insulated, covered, copper-wire 5 was wound with many (N1) layers and many (M1) windings. Likewise, the secondary coil is also a ring-type coil, in which an insulated, covered, copper-wire 5 was wound with many (N2) layers and many (M2) windings. The insulated, covered, copper-wire 5 is a general one with an insulated cover 5b such as enamel placed over a copper-wire 5a.

For example, a transformer with output 10 VA for fundamental wave with voltage 22 V has 156 as M1, 166 as M2 as well as 13 as N1 and 14 as N2, respectively.

{0021}

As shown in Figure 6, the core that forms a magnetic path between the primary coil 1 and the secondary coil 2, is a general one manufactured by stacking together of pieces of type E core and pieces of type I core of specified dimensions to specified thickness after stamping them out of an isotropic silicon steel sheet of thickness of 0.5 mm.

{0022}

The short-circuit ring 4 of conducting thin film is fabricated, as shown in Figure 5 as an example, by cutting a ring, which is approximately as thick as the primary coil 1 and the secondary coil 2, out of a rolled aluminum foil of thickness of 7  $\mu\text{m}$  and by laminating it to a durable polyester film of thickness of 50  $\mu\text{m}$ . This is fundamentally the same as the ones used in existing isolation transformers of short-circuit-ring type shown in Figure 9.

{0023}

In the first embodiment a short-circuit ring 4 of conducting thin film is positioned between each coil layer of each coil. Namely, in the primary coil 10, which is comprised by five coil layers 11, 12, 13, 14, and 15, a short-circuit ring 4 of conducting thin film is inserted in-between all the coil layers, and a short-circuit ring 4 of conducting thin film is positioned under the coil layer 11 as well as on the coil layer 15. Likewise, in the secondary coil 20, which is comprised by five coil layers 21, 22, 23, 24, and 25, a short-circuit ring 4 of conducting thin film is inserted in-between all the coil layers, and a short-circuit ring 4 of conducting thin film is positioned under the coil layer 21 as well as on the coil layer 25. Therefore, in the isolation transformer of the first embodiment there are in all twelve short-circuit rings 4 of conducting thin film used with six of them positioned in the primary coil and six others in the secondary coil.

{0024}

The method to position a short-circuit ring 4 made of a ~~planer~~planar, ring-shaped, conducting thin film into the space between coil layers in an isolation transformer of short-circuit-ring type shown in Figure 4 is as follows: Namely, at the bottom of a bobbin, which is not illustrated, the first short-circuit ring 4 made of a ~~planer~~planar, ring-shaped, conducting thin film is placed, and an insulated, covered copper-wire 5 is wound once by a winding machine to make a layer, forming a ~~planer~~planar spiral, which is not-illustrated. Then at the coil layer 11, which was just wound, the second short-circuit ring 4 made of a ~~planer~~planar, ring-shaped, conducting thin

film is placed. Likewise, at the coil layer 12, which was already wound, the third short-circuit ring 4 made of a ~~planer~~planar, ring-shaped, conducting thin film is placed. This procedure is repeated to place a short-circuit ring 4 made of a ~~planer~~planar, ring-shaped, conducting thin film in every inter-layer space, that is the space between a couple of neighboring layers, until a short-circuit ring 4 of conducting thin film is placed on the last coil layer 25.

{0025}

It is possible to ground the short-circuit rings 4 of conducting thin film positioned in-between every layer as well as the short-circuit rings 4 of conducting thin film positioned at top- and bottom surfaces of each coil. Grounding makes the short-circuit rings function as shield plates.

{0026}

In the first embodiment of the present invention, each short-circuit ring 4 of this metallic thin film with a large surface area functions in principle in the same way as former isolation transformers of short-circuit-ring type depicted in Figure 9. Namely, through this short-circuit ring 4 of conducting thin film flows the current induced by fundamental-wave current flowing through the primary coil, its higher harmonics, and some high-frequency noise from external sources. In this case, as the high-frequency components flow essentially at the surface of the conductor only owing to the skin effect, they flow around the short-circuit ring, even if the ring is thin. Therefore, being effectively attenuated by the resistivity of the short-circuit ring 4 of conducting thin film, it is hard for the high-frequency noise

to flow from the primary coil to the secondary coil. At the same time, for the resistivity of each short-circuit ring functions as a resistance commonly inserted into all of the resonance circuits, the resistivity of the short-circuit rings has the effect of dramatically reducing the amplitudes of resonances, which are distributed in the form of a number of infinitesimally small irregularly present resonance circuits due to complex combinations of capacitance and leakage inductances irregularly distributed in the coils. Meanwhile, the currents induced by the fundamental wave, which are low-frequency components, are reduced in proportion to the cross sectional area of the conducting thin film of the short-circuit ring 4. However, as the short-circuit ring is made of a thin film of thickness of  $7\ \mu\text{m}$ , its cross section is small albeit its width. Therefore, the induced currents of fundamental-wave component that flow through the short-circuit ring are extremely small. As a consequence, in isolation transformers of short-circuit ring type, which are the first embodiment, the troubles caused by high-frequency noise components could be eliminated or excluded, while keeping the loss of the fundamental-wave component negligibly small.

{0027}

Though the number of short-circuit ring 4 of conducting thin film used in existing isolation transformers of short-circuit-ring type is one, as many as twelve short-circuit rings 4 are adopted in the first embodiment of the present invention. And these numerous short-circuit rings 4 are inserted in-between every layer that forms coils. For this reason, a tertiary coil is effectively formed adjacent to each coil layer; intensifying the electromagnetic interaction between each coil and its neighboring tertiary

coil, i.e., a short-circuit ring 4 of conducting thin film. That is why the short-circuit rings 4 may eliminate or exclude the troubles caused by high-frequency noise components more effectively.

{0028}

Moreover, in the existing isolation transformers shown in Figure 4, in which a single short-circuit ring 4 is used, the distance from the short-circuit ring of conducting thin film to each coil layer is different, and therefore the effect due to conducting thin film of eliminating or excluding the troubles caused by high-frequency noise components does not reach all parts of a coil on the average. In contrast, in the isolation transformer of the first embodiment, in which a short-circuit ring 4 of conducting thin film is positioned tightly adjacent to every coil layer, it is possible for the effect of eliminating or excluding the troubles caused by high-frequency noise components due to conducting thin film to reach all parts of a coil on the average.

{0029}

For this reason, the isolation transformer of the first embodiment, in which a short-circuit ring 4 of conducting thin film is placed in each inter-coil-layer space, shows smaller amplitudes of attenuation characteristic curve with crests and troughs of various sizes more averaged and contracted than previous ones. Consequently, in comparison with the existing isolation transformers of short-circuit-ring type with a wide short-circuit ring 4 of conducting thin film covering each surface of primary- and secondary coils

or with the isolation transformers of short-circuit-ring type with a wide short-circuit ring of conducting thin film positioned between and in the vicinity of primary- and secondary coils, the isolation transformers of short-circuit-ring type in the first embodiment, have flatter characteristic curves of noise attenuation rates overall, eliminating or filtering high-frequency noise components much more effectively. In this way, the isolation transformers of short-circuit-type keep high noise-attenuation rates in the high-frequency range exceeding a few *MHz*, particularly in the range over 10*MHz*, while sufficiently suppressing each amplitude of the irregular saw-tooth wave comprised by a series of crests and troughs of various sizes.

{0030}

It is possible to implement an isolation transformer of short-circuit-ring type, which is shown in the first embodiment with a short-circuit ring 4 distributed to each inter-coil-layer space as shown in Figure 1, modifying it in various ways as shown in Figures 3 and 4.

{0031}

Depicted in Figure 3 is an isolation transformer with distributing short-circuit rings 4 of conducting thin film to selected inter-coil-layer spaces instead of each one. Namely, in the primary coil 1, which is comprised by six ~~planer~~planar, spiral, coil layers 11, 12, 13, 14, 15 and 16, a ~~planer~~planar, short-circuit ring 4 of conducting thin film is inserted into the space between coil layers 11 and 12, 13 and 14, 15 and 16, respectively.

Likewise, in the secondary coil 2, which is comprised by six ~~planer~~planar, spiral, coil layers 21, 22, 23, 24, 25 and 26, a short-circuit ring 4 of conducting thin film is inserted into the space between coil layers 21 and 22, 23 and 24, 25 and 26, respectively. Therefore, in the isolation transformer of short-circuit-ring type shown in Figure 3, there are in all six short-circuit rings 4 of conducting thin film used with three of them positioned in the primary coil and three others in the secondary coil.

{0032}

Depicted in Figure 4 is an isolation transformer to which this invention is applied with cylindrically-wound, cylindrical coil-layers instead of ~~planer~~planar, spiral coil-layers; the cylindrical coils are multi-layer, multi-winding coils that are comprised by plural, cylindrical coil layers with their inner radii being different by the diameter of the wire. Namely, in the primary coil 1, which is comprised by five cylindrical, coil layers 11, 12, 13, 14, and 15, a cylindrical, short-circuit ring 4 of conducting thin film is inserted into each inter-coil-layer space, respectively. Furthermore, a cylindrical short-circuit ring 4 of conducting thin film is placed at the inner surface of the coil layer 11 as well as at the outer surface of the coil layer 15, respectively. Likewise, in the secondary coil 2, which is comprised by five cylindrical, coil layers 21, 22, 23, 24, and 25, a cylindrical, short-circuit ring 4 of conducting thin film is inserted into each inter-coil-layer space, respectively. Furthermore, a cylindrical short-circuit ring 4 of conducting thin film is placed at the inner surface of the coil layer 21 as well as at the outer surface of the coil layer 25, respectively. Therefore, in the isolation

transformer of short-circuit-ring type shown in Figure 4, there are in all twelve cylindrical, short-circuit rings 4 of conducting thin film used with six of them positioned in the primary coil and six others in the secondary coil.

{0033}

The method to position a short-circuit ring 4 of conducting thin film into each space between coil layers in an isolation transformer of short-circuit-ring type shown in Figure 4 is as follows: Namely, at the outer surface of a bobbin [not illustrated] the first cylindrical, short-circuit ring 4 of conducting thin film is placed, and an insulated, covered copper-wire 5 is wound once cylindrically by a winding machine [not illustrated] to complete a single layer. Then at the coil layer 11, which was just wound, the second cylindrical, short-circuit ring 4 of conducting thin film is placed. Likewise, at the cylindrical coil layer 12, which was already wound, the third cylindrical, short-circuit ring 4 of conducting thin film is placed. This procedure is repeated to place a short-circuit ring 4 of conducting thin film in every inter-layer space, that is the space between a couple of neighboring layers, until the last, cylindrical, short-circuit ring 4 of conducting thin film is placed at the outer surface of the last cylindrical coil layer 25. A cylindrical, short-circuit ring 4 can be easily fabricated by winding once a conducting thin film with a specific width around a cylindrical coil layer.

{0034}

The isolation transformer of short-circuit-type shown in Figure 3, in which circular, short-circuit rings are inserted into some spaces between selected, spiral, coil-layers only, and the isolation transformers of short-

circuit-type shown in Figure 4, in which a cylindrical, short-circuit ring is inserted into each space in-between a couple of neighboring, cylindrical coil-layers, have attenuation rates of high-frequency noise almost as effective as those of the isolation transformer of short-circuit-ring type shown in Figure 1. However, the efficiency of the former transformers is somewhat less than the latter because of the smaller number of short-circuit rings used in the former transformers. Meanwhile, in comparison with the conventional isolation transformers of short-circuit-ring type or the isolation transformers of short-circuit-ring type with a wide short-circuit ring of a conducting thin film covering each surface of primary- and secondary coils, both of the isolation transformers of short-circuit-ring type shown in Figure 3 and the isolation transformers of short-circuit-ring type shown in Figure 4 have flatter overall characteristic curves of noise attenuation with each amplitude of the characteristic curves of attenuation rates averaged and contracted, eliminating or filtering high-frequency noise components much more effectively.

{0035}

In this way, the isolation transformers of short-circuit-type depicted in Figures 3 and 4 keep high noise-attenuation rates in the high-frequency range exceeding a few *MHz*, particularly in the range over *10MHz*, while sufficiently suppressing each amplitude of the characteristic curve, which is an irregular saw-tooth wave comprised by a series of crests and troughs of various sizes. Moreover, the isolation transformers shown in Figure 4 has

an advantage in the maneuverability in winding process over the ones shown in Figure 1.

{0036}

Next, the second embodiment of the present invention will be explained. Figure 7 depicts a cross section of an isolation transformer of the short-circuit-ring type, which is the second embodiment with its bobbin and core omitted and with much reduced numbers of windings and layers for easier comprehension. Figure 8 depicts a magnified part of Figure 7. As shown in Figure 8, the insulated, covered, copper-wire 6 is fabricated with a copper-wire 6a covered with an insulated cover 6b; the insulated cover 6b is further covered with a conducting thin film 6c. The thickness of this insulated, covered, copper-wire is approximately equal to or less than the skin depth of the induced current generated by the skin effect in the high-frequency range, in which resonances should be suppressed.

{0037}

This insulated, covered, copper-wire 6 is fabricated, coating metal such as aluminum by means of vacuum evaporation over a general insulated, covered, copper-wire, which has an insulated cover 6b such as enamel placed over a copper-wire 6a. In the multi-layer, multi-winding coil fabricated by winding the insulated, covered, copper-wire 6, as the surfaces of neighboring conducting thin films are tightly attached to each other, each neighboring coil layers sandwich a metallic thin film extremely tightly.

{0038}

The primary coil is a ring-type coil, in which this insulated, covered, copper-wire 6 is wound with many (N1) layers and many (M1) windings. Likewise, the secondary coil 2 is also a ring-type coil, in which an insulated, covered, copper-wire 6 is wound with many (N2) layers and many (M2) windings. For example, a transformer with output 10VA for a fundamental wave with voltage 22V has 156 as M1, 166 as M2 as well as 13 as N1 and 14 as N2, respectively.

{0039}

As shown in Figure 6, the core that forms a magnetic path between the primary coil 1 and the secondary coil 2, is a general one manufactured by stacking together of pieces of type E core and pieces of type E core of specified dimensions to specified thickness after stamping them out of an isotropic silicon steel sheet of thickness of 0.5 mm.

{0040}

In the second embodiment of the present invention, which is comprised by the primary- and secondary coils with multi-layers and multi-windings wound by insulated, covered, copper-wires 6, conducting thin films 6c at surfaces of neighboring wires closely contact each other, most effectively sandwiching each coil-layer by the conducting thin films 6c; thus, it collectively forms an isolation transformer of short-circuit-ring type, which is at least equivalent to the one manufactured by positioning a short-circuit ring of ~~planer~~planar, conducting, thin film to every space in-between

~~planer~~planar, spiral, coil layers, and simultaneously by positioning a short-circuit ring of cylindrical, conducting, thin film to every space in-between cylindrical, coil layers. In other words, the isolation transformer of short-circuit-ring type in the second embodiment of this invention, which is presented in Figure 7, is equivalent at least to the combination of the two isolation transformers of short-circuit-ring type in the second embodiment depicted in Figures 1 and 4.

{0041}

In this way, in the isolation transformer of the second embodiment, in which a short-circuit-ring of conducting thin film 6c is positioned closely adjacent to each other, the closely attached, neighboring, short-circuit rings of conducting thin film 6c collectively and simultaneously form ~~planer~~planar, short-circuit rings of conducting thin film with wide surface areas inserted in-between every ~~planer~~planar, coil layer as well as cylindrical short-circuit rings of conducting thin film with wide surface areas inserted in-between every cylindrical coil-layer. Subsequently, as the electromagnetic couplings between each coil layer and neighboring short-circuit rings of conducting thin film become the strongest, elimination or exclusion of the troubles caused by high-frequency noise components can be performed more effectively by short-circuit rings of conducting thin film than in the first embodiment. Moreover, as it is equivalent to positioning each short-circuit ring of conducting thin film at the closest possible distance from each coil layer, in this isolation transformers of short-circuit-ring type

the elimination or exclusion effects due to these equivalent short-circuit rings reach to every part of coils in a more averaged manner.

{0042}

Consequently, in the second embodiment each amplitude of noise attenuation characteristic curve with a series of crests and troughs of various sizes is further averaged and contracted in comparison with the first embodiment. As a result, in comparison with the conventional isolation transformers of short-circuit-ring type with a wide short-circuit ring of conducting thin film covering each surface of primary- and secondary coils and of the conventional isolation transformers of short-circuit-ring type with a wide short-circuit ring of conducting thin film positioned between and in the close vicinity of primary- and secondary coils the isolation transformers of short-circuit-ring type in the second embodiment have larger noise-attenuation rates in the high-frequency range exceeding a few *MHz*, particularly in the range over *10MHz*, while sufficiently suppressing each amplitude of the irregular saw-tooth wave comprised by a series of crests and troughs of various sizes.

{0043}

The first and second examples in which the present invention were implemented to isolation transformers comprised by a multi-layer, multi-winding primary coil, a multi-layer, multi-winding secondary coil, and a magnetic path of each coil have been explained in detail above. However, this invention is of course not limited to these examples of implementation.

Though the short-circuit rings of conducting thin film are applied to both of the multi-layer, multi-winding primary- and secondary coils in the above examples, they may be applied to either one. Similarly, though the insulated, covered, copper-wires made by coating some copper-wires with insulated films, and further coating the insulated surfaces with conducting thin films are used in both of the multi-layer, multi-winding primary- and secondary coils in the above examples, they may be used in either one.

{0044}

The configuration of the wound-up, primary- and secondary coils are not limited to circular nor rectangular ones, and coils may take any other geometry. The core that forms a magnetic path between the primary- and secondary coil is not limited to the ones manufactured by stacking together pieces of type E core and pieces of type I core shown in Figure 6, and may be a cut core or other types of cores. The isolation transformers related to the present invention may eliminate or filter high-frequency noise components even if they are used in combination with ordinary means for filtering.

{0045}

#### ~~{EFFECTS OF THE INVENTION}~~

Owing to the present invention the isolation transformers of short-circuit-type with high noise attenuation rates in the high-frequency range exceeding a few *MHz*, and particularly above 10*MHz*, while sufficiently suppressing each amplitude of the irregular, saw-tooth shaped waves with a series of crests and troughs of various sizes have been provided.

Namely, the isolation transformers of short-circuit-ring type now have flatter overall characteristic curves of noise attenuation with each amplitude of the characteristic curves of attenuation rates averaged and contracted, eliminating or filtering high-frequency noise components much more effectively in the high-frequency range exceeding a few *MHz*, and particularly above *10MHz*.

Consequently, isolation transformers with extremely high reliability have been provided in comparison with the conventional isolation transformers of short-circuit-ring type with a wide short-circuit ring of conducting thin film covering each surface of primary- and secondary coils and of the conventional isolation transformers of short-circuit-ring type with a wide short-circuit ring of a conducting thin film positioned between and in the vicinity of primary- and secondary coils.

#### ~~[BRIEF EXPLANATION OF FIGURES]~~

~~Figure 1 is a cross section view of the isolation transformer in the first embodiment.~~

~~Figure 2 is a cross section view of magnified part of the isolation transformer in the first embodiment shown in FIGURE 1.~~

~~Figure 3 is a cross section view of the isolation transformer of a modification in the first embodiment~~

~~Figure 4 is a cross section of the isolation transformer of another modification in the first embodiment~~

~~Figure 5 is a plan view of an example of circular, short circuit ring of conducting thin film~~

~~Figure 6 is a oblique view of an example of the core.~~

~~Figure 7 is a cross section of the isolation transformer in the second embodiment of the present invention~~

~~Figure 8 is a cross section view of magnified part of the isolation transformer in the second embodiment shown in FIGURE 7.~~

~~Figure 9 is a cross section of an example of existing isolation transformer of short circuit ring.~~

~~Figure 10 is a cross section of another example of conventional isolation transformer of short circuit ring type.~~

~~Figure 11 shows a characteristic curve of attenuation rate for normal mode noise of a conventional isolation transformer of short circuit ring type.~~

~~Figure 12 shows a characteristic curve of attenuation rate for normal mode noise of a conventional isolation transformer of electromagnetic shield type.~~

#### ~~[EXPLANATION OF SYMBOLS]~~

1——primary coil

2——secondary coil

3——core

4——short circuit ring of conducting thin film

5——insulated, covered, copper wire

5a-copper wire

5b-insulated film

6—insulated, covered, copper wire

6a-copper wire

6b-insulated film

6c-conducting thin film

11~16 coil layers of a primary coil

21~26 coil layers of a secondary coil

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